

REVISION CLASS 1

Aidan Hogg - 21 November 2019
ELEC96010 (EE3-07): Digital Signal Processing
Department of Electrical and Electronic Engineering

Method:

- 1: Conceptual question posed - students think quietly on their own and report initial answers on Mentimeter (3 mins)
- 2 Students discuss their answers in small groups (2 mins)
- 3 Explanation/discussion of correct answer (3 mins)

Consider the following statements:

- 1: Allpass filters have mirror image numerator and denominator coefficients
- 2 In an allpass filter, the zeros are the poles reflected in the unit circle
- 3 Allpass filters have a gain magnitude of 1 even with coefficient errors

Which of these statements are true?

- A: 1
- B: 1 and 2
- C: All 3 of them

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Allpass filters have mirror image numerator and denominator coefficients

$$[] = [-] \quad () = - (-1)$$

$$() = \frac{=0 []^-}{=0 []^- (-)} = \frac{=0 []^-}{=0 []}$$

$$| () | = 1$$

Allpass filters have a gain magnitude of 1 even with coefficient errors

First Order:

$$() = \frac{- + -1}{1 - -1} = - \frac{1 - -1 -1}{1 - -1}$$

Poles at and zero at -1 : 'reflected in unit circle'

Constant distance ratio $- = | | - \frac{1}{2}$

Consider the following system:

$$[] = 2 [] - 3 [- 1] + [- 2]$$

Where are the poles and zeros located?

A: Zeros at = $\{2, 0\}$ and $\{1, 0\}$ Poles at = $\{0, 0\} \times 2$

B: Zeros at = $\{2, 0\}$ and $\{1, 0\}$ Pole at = $\{0, 0\}$

C: Zeros at = $\{1/2, 0\}$ and $\{1, 0\}$ Poles at = $\{0, 0\} \times 2$

Consider the following system:

$$[\] = 2 [\] - 3 [\ - 1] + [\ - 2]$$

Where are the poles and zeros located?

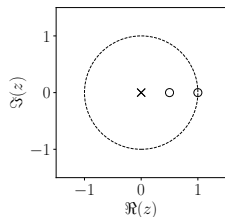
A: Zeros at $= \{2, 0\}$ and $\{1, 0\}$ Poles at $= \{0, 0\} \times 2$

B: Zeros at $= \{2, 0\}$ and $\{1, 0\}$ Pole at $= \{0, 0\}$

C: Zeros at $= \{1/2, 0\}$ and $\{1, 0\}$ Poles at $= \{0, 0\} \times 2$

Solution

$$\begin{aligned}
 [] &= 2 [] - 3 [- 1] + [- 2] \\
 () &= 2 () - 3^{-1} () +^{-2} () \\
 () &= [^{-2} - 3^{-1} + 2] () \\
 () &= [(^{-1} - 2)(^{-1} - 1)] ()
 \end{aligned}$$



Therefore

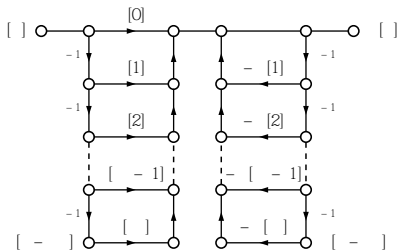
$$\text{Zeros at } = \left\{ \frac{1}{2}, 0 \right\} \text{ and } \{ 1, 0 \}$$

$$\text{Poles at } = \{ 0, 0 \} \text{ and } \{ 0, 0 \}$$

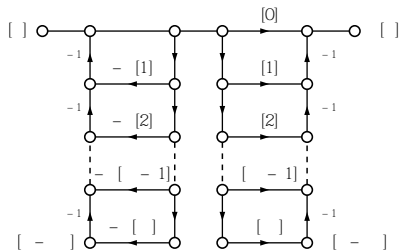
Consider the filter:

$$Y(z) = \frac{[0] + [1]z^{-1} + []z^{-2}}{1 + [1]z^{-1} + []z^{-2}}$$

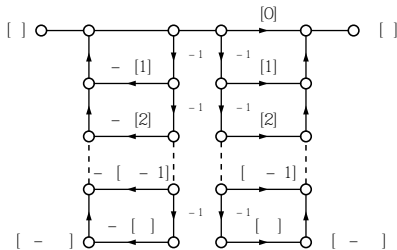
Which digital filter structure is implementing Direct Form I Transposed?



A



B

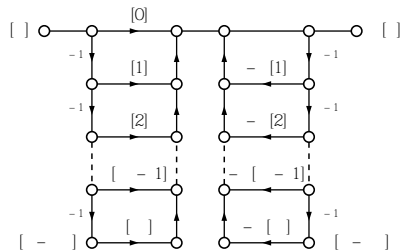


C

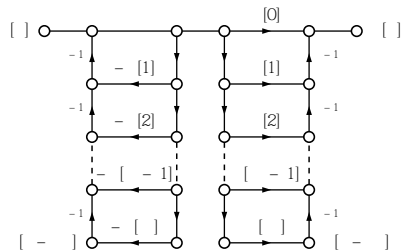
Consider the filter:

$$Y(z) = \frac{[0] + [1]z^{-1} + []z^{-2}}{1 + [1]z^{-1} + []z^{-2}}$$

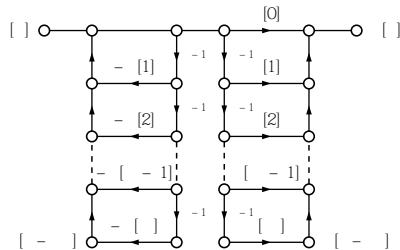
Which digital filter structure is implementing Direct Form I Transposed?



A



B



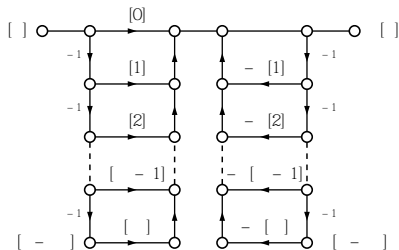
C

EXPLANATION

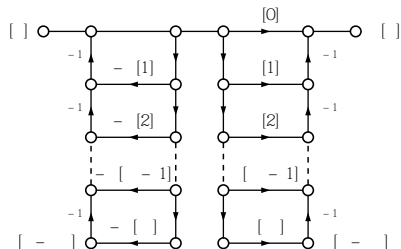
Consider the filter:

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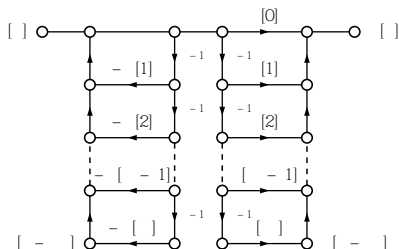
Which digital filter structure is implementing Direct Form I Transposed?



Direct Form I



Direct Form I Transposed



Direct Form II

If an IIR filter has a transfer function:

$$H(z) = \frac{N(z)}{D(z)} = \frac{[0] + [1]z^{-1} + [2]z^{-2} + \dots + [M-1]z^{-(M-1)} + [M]z^{-M}}{1 + [1]z^{-1} + [2]z^{-2} + \dots + [N-1]z^{-(N-1)} + [N]z^{-N}}$$

Then direct forms use coefficients $[a_k]$ and $[b_k]$ directly. This can be implemented as a cascade of two filter sections where:

$$H_1(z) = \frac{N(z)}{D(z)} = [0] + [1]z^{-1} + [2]z^{-2} + \dots + [M-1]z^{-(M-1)} + [M]z^{-M}$$

$$H_2(z) = \frac{1}{D(z)} = \frac{1}{1 + [1]z^{-1} + [2]z^{-2} + \dots + [N-1]z^{-(N-1)} + [N]z^{-N}}$$

Note that $H_1(z)$ can be seen as an FIR filter and the time-domain representation of $H_2(z)$ is given by:

$$[1] = [0] - [1][z^{-1}] - [2][z^{-2}] - \dots - [N-1][z^{-(N-1)}] - [N][z^{-N}]$$

Direct form I can be viewed as $H_1(z)$ followed by $\frac{1}{D(z)}$.

Direct form II implements $\frac{1}{()}$ followed by $()$

Transposed Forms

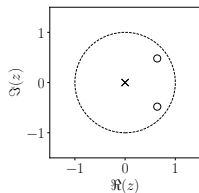
It is also possible to convert any structure into an equivalent transposed form. This is achieved in the following way:

- 1: Reverse direction of each interconnection
- 2: Reverse direction of each multiplier
- 3: Change junctions to adders and vice-versa
- 4: Interchange the input and output signals

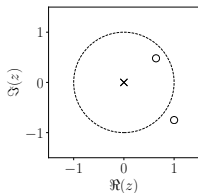
Check: A valid structure must never have any feedback loops that don't go through a delay (z^{-1} block).

An FIR filter () is determined by the zeros of

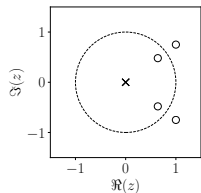
$$H(z) = \prod_{k=1}^N (z - z_k)$$



1



2



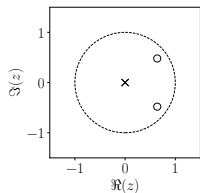
3

Which FIR filter has symmetric coefficients []?

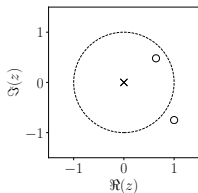
- A: 1 and 3
- B: 2 and 3
- C: 3

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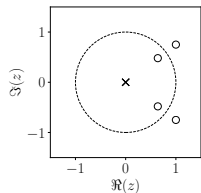
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1



2



3

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Symmetric properties:

Real [] conjugate zero pairs:

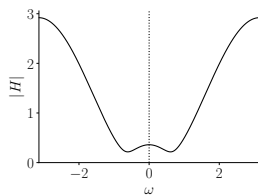
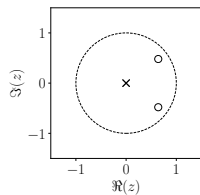
Symmetric: [] = [-] reciprocal zero pairs: -1

Real & Symmetric [] conjugate and reciprocal groups of four (else pairs on the real axis)

EXPLANATION

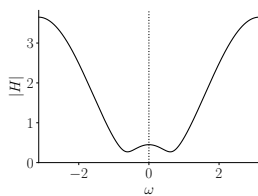
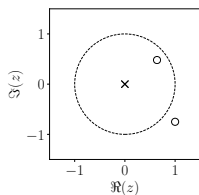
Real

[1, -1.28, 0.64]



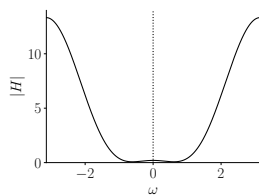
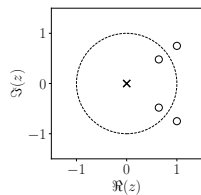
Symmetric

[1, -1.64 + 0.27j, 1]



Real & Symmetric

[1, -3.28, 4.7625, -3.28, 1]



In all of the proofs below, we assume that $\lambda = \lambda_0$ is a root of $P(\lambda)$ so that $P(\lambda_0) = \sum_{k=0}^n a_k \lambda_0^k = 0$ and then we prove that this implies that other values of λ also satisfy $P(\lambda) = 0$.

(1) Real λ

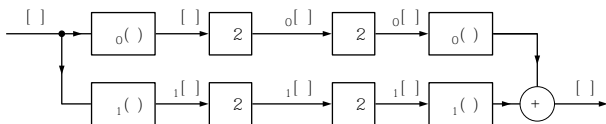
$$\begin{aligned}
 P(\lambda) &= \sum_{k=0}^n a_k (\lambda_0)^k \\
 &= \sum_{k=0}^n a_k (\lambda_0)^k \quad \text{since } a_k \text{ is real} \\
 &= \left(\sum_{k=0}^n a_k \lambda_0^k \right) \quad \text{take complex conjugate} \\
 &= 0 = 0 \quad \text{since } P(\lambda_0) = 0
 \end{aligned}$$

In all of the proofs below, we assume that $\lambda = \lambda_0$ is a root of $\chi(\lambda)$ so that $\chi(\lambda_0) = \det(A - \lambda_0 I) = 0$ and then we prove that this implies that other values of λ also satisfy $\chi(\lambda) = 0$.

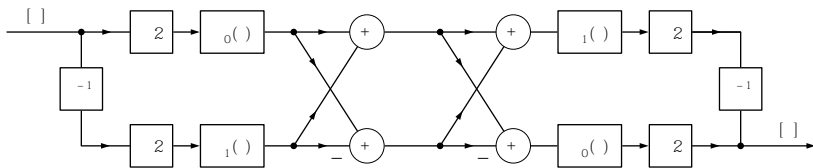
(2) Symmetric: $A = A^T$

$$\begin{aligned}
 \chi(\lambda_0^{-1}) &= \det(A - \lambda_0^{-1} I) \\
 &= \det(A - \lambda_0^{-1} I)^T && \text{substitute } \lambda = \lambda_0^{-1} \\
 &= \det(A^T - \lambda_0^{-1} I) && \text{take out } \lambda_0^{-1} \text{ factor} \\
 &= \det(A - \lambda_0^{-1} I) && \text{since } A^T = A \\
 &= 0 \times 0 = 0 && \text{since } \chi(\lambda_0) = 0
 \end{aligned}$$

Consider the following Quadrature Mirror Filterbank (QMF):



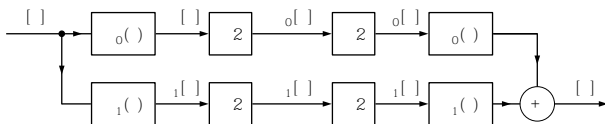
Which can be implemented using polyphase QMF:



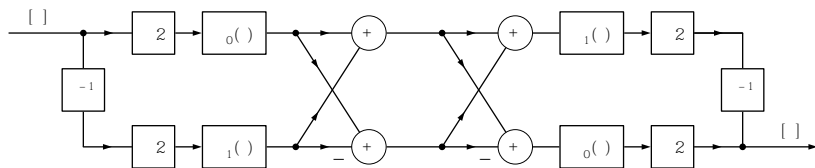
What is the computational saving achieved by the polyphase QMF?

- A: A factor of 2
- B: A factor of 4
- C: A factor of 8

Consider the following Quadrature Mirror Filterbank (QMF):



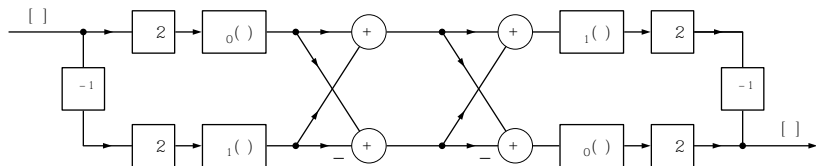
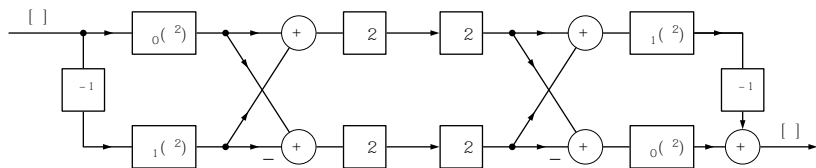
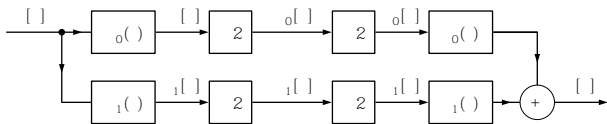
Which can be implemented using polyphase QMF:



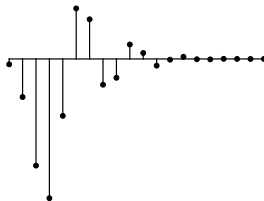
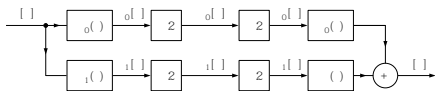
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EXPLANATION

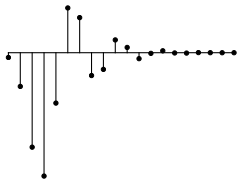


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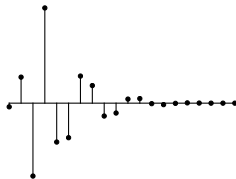


Impulse response $h_0[n]$

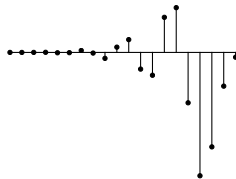
Which is the correct impulse response for $h_1[n]$?



A

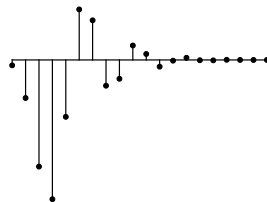
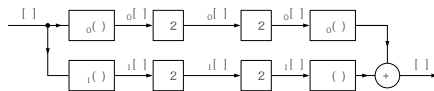


B



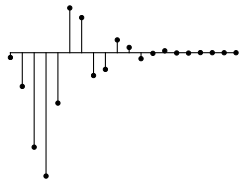
C

Consider the following Quadrature Mirror Filterbank (QMF):

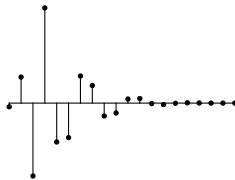


Impulse response $h_0[n]$

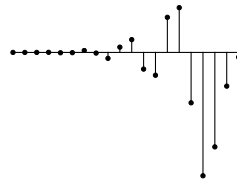
Which is the correct impulse response for $h_1[n]$?



A

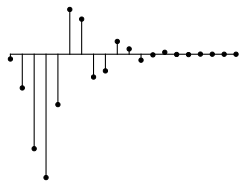
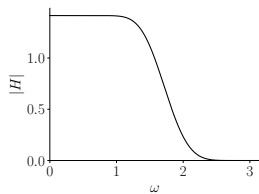


B

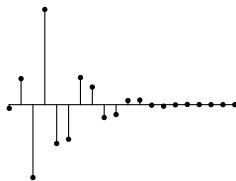
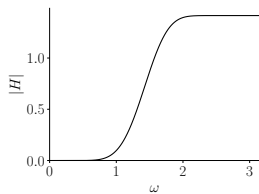


C

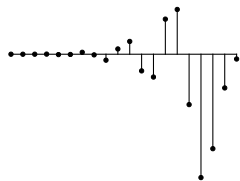
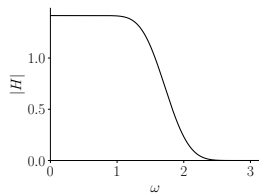
Magnitude responses:



$$h_1[n] = o[n]$$

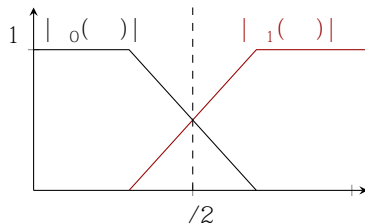


$$h_1[n] = (-1)^n o[n]$$



$$h_1[n] = o[-n - 1]$$

Proof:



The relationship between $h_0(\omega)$ and $h_1(\omega)$ when they are quadrature mirror filters is $h_1(\omega) = h_0(-\omega)$.

The corresponding relationship between the impulse responses of these filters is $h_1[n] = (-1)^n h_0[n]$.

$$h_1(\omega) = h_0(-\omega) = \sum_{n=-\infty}^{\infty} h_0[n](-1)^n e^{j\omega n} = \sum_{n=-\infty}^{\infty} h_0[n](-1)^n e^{-j\omega n}$$

There will be a class next Thursday (Week 9 - 28/11/2019)

Room: 403A/B

Time: 15:00 to 16:00